# Simulation and Design of A Single Phase Inverter with Digital PWM Issued by An Arduino Board

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Abstract— The current paper has as major purpose the design of a single-phase inverter for educational purposes. This project has the aim to use Arduino board to ease the Pulse Width Modulation (PWM) implementation on a single-phase inverter, substituting analogical circuitry. To achieve those aims, a first complete theoretical analysis will be made, including the study of the different conventional PWM techniques. The complete design is modeled in Proteus software and its output is verified practically.

Keywords— Single-phase inverter, PWM, Arduino; Proteus simulations

## I. INTRODUCTION

Inverters are circuits that transfer power from a DC source to an AC load. Inverters are used in applications such as adjustable-speed ac motor drives, uninterruptible power supplies (UPS), active filters, flexible ac transmission systems (FACTS), running ac appliances from an automobile battery...

Inverters can be classified into many types based on output, source, type of load...

Complete classification of inverter circuits is as follows:

TABLE I. classification of inverter circuits

Output	Source	Type of Load
-Square Wave	-Current Source	-Single Phase
-Sine Wave	-Voltage Source	-Three Phase
	_	

There are several control techniques for inverters. The most common one is the Pulse Width Modulation (PWM) technique. The main aim of these modulation techniques is to enhance the output of the inverters by obtaining an output voltage or current very close to sine waveform. PWM technique provides a way to push harmonics to higher frequencies, making filtering easier. There are many types of modulation techniques, the most familiar are:

- Sinusoidal Pulse Width Modulation (SPWM)
- Modified Pulse Width Modulation (MPWM)
- Random Pulse Width Modulation (RPWM)
- Space Vector Modulation (SVM)
- Delta Modulation (DM)

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## - Specific Harmonic Elimination (SHE)

The main objective of this project is the design, simulation and testing of a single-phase inverter for educational purposes. In order to achieve this, the first step is a theoretical reminder about inverters. The most important task in this project is the implementation of the PWM modulation digitally. For this purpose, the way to obtain the PWM signal from the Arduino board is explained. After that, Proteus simulations are carried out in order to have a better idea about the results expected. Once the theoretical simulation is made, the circuit should be built into a protoboard for real testing.

## II. THEORETICAL BACKGROUND

Inverters are circuits that convert DC to AC. The full-bridge converter of Fig.1 is the basic structure of an inverter.

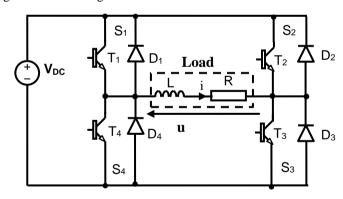


Fig.1: the full-bridge inverter

The switches  $S_1$ ,  $S_2$ ,  $S_3$ ,  $S_4$  in the full-bridge inverter must be capable of carrying both positive and negative currents. Therefore, a feedback diode is placed in parallel (antiparallel) with each switch.

The semiconductor devices  $T_1$ ,  $T_2$ ,  $T_3$ ,  $T_4$  must be fully controllable switches which can be turned on and off (such as BJT, MOSFET, JFET, IGBT, GTO, MCT...).

## A. The square-wave inverter

The simplest switching scheme for the full-bridge converter produces a square wave output voltage. The

switches connect the load to  $V_{DC}$  for first half cycle 0 < t < T/2when  $S_1$  and  $S_3$  are closed, or to  $-V_{DC}$  for second half cycle T/2 < t < T when  $S_2$  and  $S_4$  are closed. The periodic switching of the load voltage between  $V_{DC}$  and  $-V_{DC}$  produces a square wave voltage across the load.

The current waveform in the load depends on the load components. For a resistive load, the current has the same shape of the output voltage. An inductive load will have a current closer to a sinusoidal form than the voltage because of the filtering property of the inductance.

In order to have the exact expression of current for a series RL load, the following two equations must be solved:

For 
$$\in \left[0, \frac{T}{2}\right]$$
:  $S_1$  and  $S_3$  are closed:

$$u = V_{DC} = R \ i + L \frac{di}{dt} \tag{1}$$

 $u = V_{DC} = R \ i + L \frac{di}{dt}$  For  $\in \left[\frac{T}{2}, T\right]$ : S<sub>2</sub> and S<sub>4</sub> are closed:

$$u = -V_{DC} = R \ i + L \frac{di}{dt} \tag{2}$$

It can be shown that:

For 
$$t \in \left[0, \frac{T}{2}\right]$$
:

$$i(t) = \frac{V_{DC}}{R} + \left(I_{min} - \frac{V_{DC}}{R}\right) e^{-\frac{t}{\tau}}$$

$$\text{For } t \in \left[\frac{T}{2}, T\right]:$$
(3)

$$i(t) = -\frac{V_{DC}}{R} + \left(I_{max} + \frac{V_{DC}}{R}\right)e^{-\frac{(t-T/2)}{\tau}} \tag{4}$$

Where 
$$\tau = \frac{L}{R}$$
 and  $I_{max} = -I_{min} = \frac{V_{DC}}{R} \left( \frac{1 - e^{-T/2\tau}}{1 + e^{-T/2\tau}} \right)$   
The square wave output voltage u(t) and the steady-state

current waveform i(t) for an R-L load are shown in Fig.2:

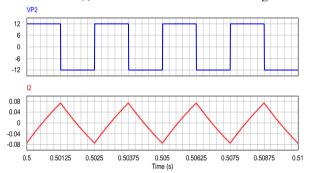


Fig.2: output voltage and current for an R-L load (V<sub>DC</sub>=12V, R=25Ω, L=100mH, f=400Hz)

Since the objective of the inverter is to supply a load with an AC current, it is useful to describe the quality of the ac output voltage and current: a Fourier analysis is necessary. The output voltage u(t) and load current i(t) must be expressed in terms of a Fourier series.

In the case of the square wave u(t), the Fourier series contains the odd harmonics and can be represented as:

$$u(t) = \sum_{n \, (odd)} \frac{4 \, V_{DC}}{n\pi} \sin \left(2n\pi f t\right) \tag{5}$$

For the load current i(t), the Fourier series is as follows:

$$i(t) = \sum_{n \text{ (odd)}} I_n \sin(2n\pi f t + \varphi_n)$$
 (6)

The amplitude of each current term is given by:
$$I_n = \frac{\frac{4V_{DC}}{n\pi}}{|Z_n|} = \frac{\frac{4V_{DC}}{n\pi}}{\sqrt{R^2 + (2n\pi fL)^2}}$$
(7)

As the harmonic number n increases, the amplitude of the Fourier voltage component decreases and the magnitude of the

corresponding impedance increases, both resulting in small currents for higher-order harmonics as shown in Fig.3:

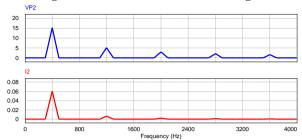


Fig.3: output voltage and current spectrum

The quality of a non-sinusoidal wave can be expressed in terms of total harmonic distortion (THD). The closer the waveform to sinusoidal, the smaller is the THD.

For square wave inverter with  $(V_{DC}=12V, R=25\Omega,$ L=100mH, f=400Hz), the THD output voltage and the THD load current are respectively:

$$THD_u = 48.3\% \tag{8}$$

$$THD_i = 12.2\%$$
 (9)

In conclusion, for square wave inverter, the first harmonics are very close to the fundamental which makes filtering difficult. On another hand, the THD level proves that the quality of the output waveforms had to be improved.

#### B. PWM inverter

Pulse-width modulation (PWM) provides a way to decrease the total harmonic distortion of load current: the harmonics will be at much higher frequencies than for a square wave, making filtering easier.

There are many types of modulation techniques. In this project, two types of PWM will be used:

- Sinusoidal Pulse Width Modulation (SPWM)
- Specific Harmonic Elimination (SHE)

#### 1) Sinusoidal Pulse Width Modulation (SPWM)

In the sinusoidal pulse width modulation (SPWM), the gate control signals are generated by comparing a sinusoidal reference voltage signal (vsine) with a high-frequency triangular carrier voltage signal (vtri). The intersection points of the sinusoidal reference voltage signal and the triangular carrier voltage signal determine the turn on and turn off instants of the switching devices.

When  $v_{sine} > v_{tri}$ :  $S_1$  and  $S_3$  are closed then:  $u = V_{DC}$ When  $v_{sine} < v_{tri}$ :  $S_2$  and  $S_4$  are closed then:  $u = -V_{DC}$ This version of PWM is bipolar because the output alternates between  $+V_{DC}$  and  $-V_{DC}$ .

Fig.4 illustrates the principle of sinusoidal bipolar pulsewidth modulation.

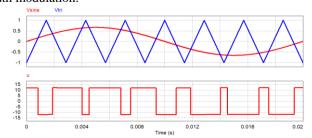


Fig.4: principle of bipolar SPWM

Two important parameters characterize the SPWM:

The frequency modulation ratio  $\mathbf{m_f}$ , defined as the ratio of the frequencies of the carrier and reference signals:

$$m_f = \frac{f_{tri}}{f_{sine}} \tag{10}$$

The amplitude modulation ratio  $m_a$ , defined as the ratio of the amplitudes of the reference and carrier signals:

$$m_a = \frac{V_{m,sine}}{V_{m,twi}} \tag{11}$$

 $m_a = \frac{v_{m,sine}}{v_{m,tri}}$  (11) Simulation with PSIM for the following values:  $f_{sine} =$ 50Hz,  $m_f = 7$ ,  $m_a = \frac{2}{3}$  gives the following waveforms for output voltage and current:

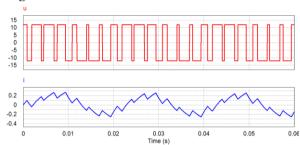


Fig.5: output voltage and current for SPWM ( $f_{sine} = 50Hz$ ,  $m_f = 7, m_a = \frac{2}{3}$ 

The output voltage and current spectrum are given in Fig.6:

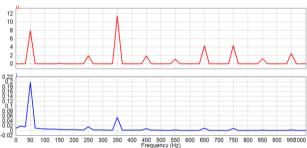


Fig.6: output voltage and current spectrum for SPWM

Fig.6 shows that thanks to SPWM technique, the first considerable harmonic is at 350Hz which corresponds to  $m_f \times f_{sine}$ .

On another hand, the THD load current is:

$$THD_i = 28.3\%$$
 (12)

The unfiltered PWM output current has a relatively high THD, but the harmonics are at much higher frequencies than for a square wave, making filtering easier.

## 2) Specific Harmonic Elimination (SHE)

The specific harmonic elimination modulation consists in determining a set of switching angles in order to eliminate certain harmonics in the PWM inverter output.

The output voltage for bipolar PWM has the following form:

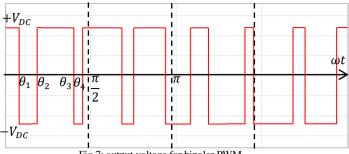


Fig.7: output voltage for bipolar PWM

The rms value of the fundamental and the harmonics for the ouput voltage u are given by:

$$U_n = \frac{4\sqrt{2} v_{DC}}{n \pi} \left[ \frac{1}{2} - \cos n\theta_1 + \cos n\theta_2 - \cos n\theta_3 + \cos n\theta_4 \right]$$
(13)

With four angles  $(\theta_1, \theta_2, \theta_3 \text{ and } \theta_4)$ , four harmonics can be eliminated. Logically, harmonics 3, 5, 7 and 9 must be eliminated. To find the adequate angles the following system of equations must be solved:

$$\begin{cases} \frac{1}{2} - \cos 3\theta_1 + \cos 3\theta_2 - \cos 3\theta_3 + \cos 3\theta_4 = 0\\ \frac{1}{2} - \cos 5\theta_1 + \cos 5\theta_2 - \cos 5\theta_3 + \cos 5\theta_4 = 0\\ \frac{1}{2} - \cos 7\theta_1 + \cos 7\theta_2 - \cos 7\theta_3 + \cos 7\theta_4 = 0\\ \frac{1}{2} - \cos 9\theta_1 + \cos 9\theta_2 - \cos 9\theta_3 + \cos 9\theta_4 = 0 \end{cases}$$
(14)

The angles are found by means of iterative algorithms as no analytical solutions can be derived. This task is performed by the **Matlab** software as following:



Fig.8: Matlab code used to find switching angles

The code gives the following result:



Fig.9: the four switching angles

These angles will be used later in the Arduino code.

## ARCHITECTURE OF THE SINGLE PHASE **INVERTER**

The block diagram of the whole circuit is shown in Fig.10:

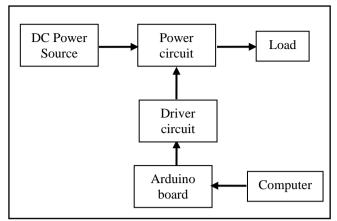


Fig.10: Block Diagram of Single Phase Inverter

The block diagram has three modules:

- Power Circuit Module
- Control Circuit Module
- Driver Circuit Module

#### A. Power circuit module:

The power circuit consists of the full-bridge inverter shown in Fig.1. The input to the power circuit is a DC source  $(V_{DC} = 12V)$ . The full-bridge contains four MOSFET (IRF530) and four antiparallel diodes (1N4007).

The load is an R-L series circuit with  $R = 25\Omega$  and L =100mH.

#### B. Arduino board:

For this project an Arduino UNO will be selected. Widely employed in basic programming projects, it contains a 16MHz clock, 14 digital input/output pins, 6 analog inputs and 6 PWM outputs. It is based on Atmega 328P microcontroller and can be powered via USB connection or by an external power supply.



Fig.11: Arduino UNO board

The main advantages of Arduino are:

- Simple and clear programming environment.
- Inexpensive.
- Open source and extensible software and hardware.
- Wide range of tutorials and information available.

Thus, it seems to be the perfect choice for educational purposes, allowing students to take their first steps in programming.

The Arduino UNO board is used to produce the PWM pulse. This PWM pulse oscillates between 0 and 5 Volts. In order to have the appropriate voltage level on MOSFET gates, the pulse obtained from Arduino board must be given to the driver circuit.

#### C. Driver circuit module:

Generally micro controllers generate PWM signals of very low range which is not sufficient for the switches to turn on so a gate driver circuit is designed for that purpose. In this paper, **IR2110** is used for this purpose.

As explained on its datasheet, it is a high voltage, high speed MOSFET and IGBT driver with independent high and low side output channels.

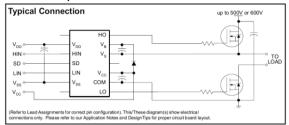


Fig.12: Typical connection of IR2110

The list of components used is given in the following table.

TABLE II. List of components used

4 x MOSFET	IRF540
2 x Driver	IR2110
6 x Diode	1N4007
4 x Resistor	$47\Omega$
4 x Resistor	1kΩ
2 x Capacitor	22μF/50V
2 x Capacitor	100nF

#### PROTEUS SIMULATIONS

In this section, Proteus simulations are carried out for the square wave and the PWM inverters (SPWM+SHE).

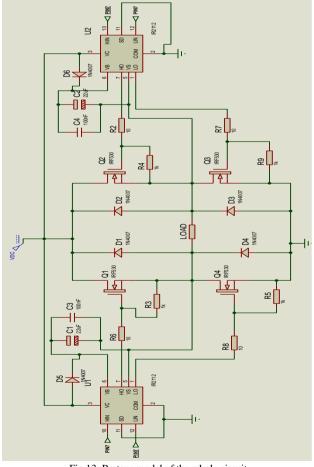


Fig.13: Proteus model of the whole circuit

#### A. Square wave inverter:

PIN7 of the Arduino board deliver a pulse that oscillates between 0 and 5V. The following code must be written with the Arduino Software (IDE).



Fig.14: Arduino code for square-wave inverter

Simulation results for the square-wave inverter are given in Fig.15. The frequency is f = 400Hz, the load values are  $R = 25\Omega$  and L = 100mH.

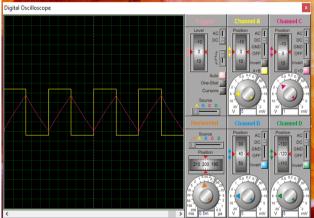


Fig.15: Proteus simulation for square wave inverter

The red and yellow waveforms represent respectively the load current and the output voltage.

## B. SPWM inverter:

As explained previously, for the SPWM inverter, the turn on and turn off instants of the switching devices are determined from the intersection points of a sinusoidal reference voltage signal and a triangular carrier voltage signal. These intersection points are determined by MATLAB.

As a first step, a MATLAB code (Fig.16, Fig.17) permits to plot both sinusoidal and triangular signals and to find the abscissas of the intersection points of the two signals. The parameters of the SPWM are:  $f_{sine} = 50Hz$ ,  $m_f = 7$ ,  $m_a = \frac{2}{3}$ 

```
Editor - C:\Users\ymoez\Desktop\moez\acada\projets\projet-conv
   M.m
         ×
 1
         t=0:0.00001:0.02;
         f=50;
 2
 3
 4
         r=2/3;
 5
         y= sawtooth(2*pi*m*f*t,1/2);
 6 -
         x=r*sin(2*pi*f*t);
         [xi, yi] = polyxpoly(t, x, t, y);
 8
         plot(t,y,t,x,xi,yi,'b*')
 9
```

Fig.16: MATLAB code to determine turn on and turn off instants of the switching devices

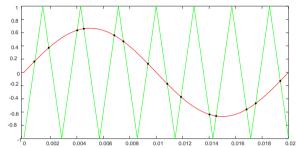


Fig.17: sinusoidal and triangular signals and their intersection points

The abscissas of the intersection points of the two signals given by MATLAB are:

{0.0008; 0.0019; 0.0040; 0.0045; 0.0068; 0.0075; 0.0094; 0.0108; 0.0119; 0.0140; 0.0145; 0.0168; 0.0175; 0.0194}

These values are injected in the arduino code in order to generate the SPWM signal. The following code must be written with the Arduino Software (IDE).



Fig.18: Arduino code for SPWM inverter

Simulation results for SPWM inverter are given in Fig.19.

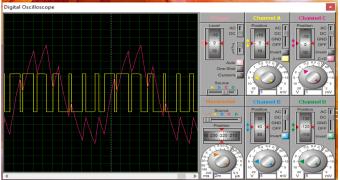


Fig.19: Proteus simulation for SPWM inverter

The red and yellow waveforms represent respectively the load current and the output voltage.

Fig.20 shows output voltage (green) and current (red) spectrum.



Fig.20: output voltage and current spectrum for SPWM

Fig.20 shows that the first considerable harmonic is at 350Hz which corresponds to  $m_f \times f_{sine}$ .

## C. Specific Harmonic Elimination modulation (SHE):

As explained previously, the specific harmonic elimination modulation consists in determining a set of switching angles in order to eliminate certain harmonics in the PWM inverter output. These angles are obtained from Matlab by resolving the system (14). The switching angles (shown in Fig.9) are injected in the arduino code in order to generate the SHE modulation signal. The following code must be written with the Arduino Software (IDE).



Fig.21: Arduino code for SHE inverter

Simulation results for SHE modulation inverter are given in Fig.22.

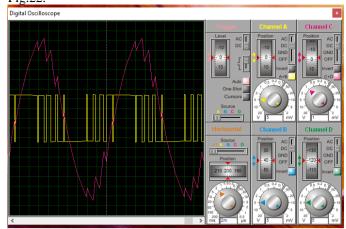


Fig.22: Proteus simulation for SHE inverter

The red and yellow waveforms represent respectively the load current and the output voltage.

Fig.23 shows output voltage (green) and current (red) spectrum (FFT).

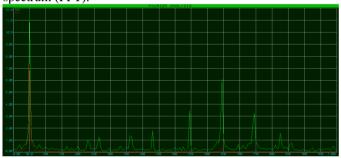
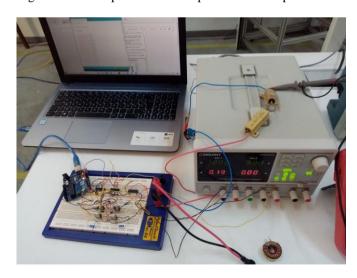


Fig.23: output voltage and current spectrum for SHE PWM Fig.23 shows that the first considerable harmonic is at 550Hz which corresponds to  $11 \times f_{sine}$ . Harmonics 3,5,7,9 are quasi eliminated.

## V. EXPERIMENTAL RESULTS

Fig.24 shows the practical circuit performed on a protoboard.



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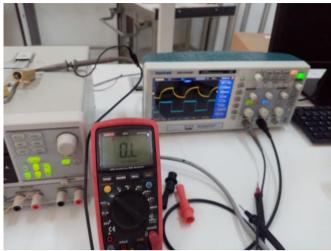


Fig.24: complete circuit setup

Fig.25, Fig.26 and Fig.27 show practical results respectively for square wave inverter, SPWM inverter and specific Harmonic Elimination modulation inverter. Yellow waveform represents the load current. The blue waveform represents the output voltage.

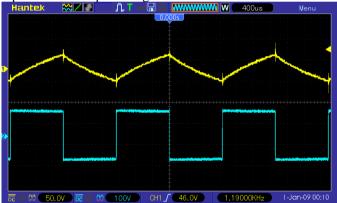


Fig.25: practical results for square wave inverter

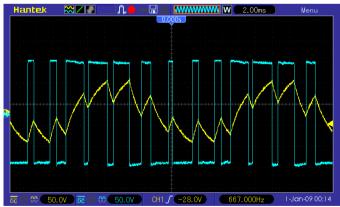


Fig.26: practical results for SPWM inverter

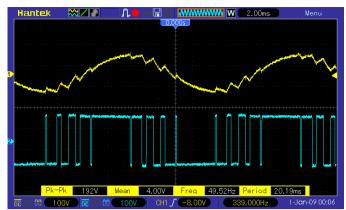


Fig.27: practical results for the specific Harmonic Elimination modulation inverter

Practical measurements confirm Proteus simulation for all the studied cases.

## VI. CONCLUSION

In this project, a single phase inverter is designed. For this inverter, the pulses needed for switching semiconductor devices are generated using the Arduino Uno board and the IR2110 driver.

The most important task in this project is the implementation of the PWM modulation on the Arduino board. Two types of PWM modulation are studied: sinusoidal Pulse Width Modulation (SPWM) and Specific Harmonic Elimination (SHE).

Before performing experimental measurements, simulations with the Proteus software are carried out for the square wave and the PWM inverters (SPWM+SHE). The simulations results obtained are in accordance with measurements.

Through this project, the most important conclusion is that Arduino offers a simple and efficient programming environment suitable for PWM converters.

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