

initializes the timers and I/O ports then reads the commands from the keyboard. Each key is designated with a specific count which is equivalent to different duty cycles. On the basis of the count Microcontroller generates a PWM signal and the motor can be stopped by keying a specific character anytime.

In this project all of three timers are used:

·**Timer 0** in Normal Mode was used to count pulses from motor speed encoder.

·**Timer 1**, which has two independent PWM channels, was used to generate PWM waveform to control of the motor speed, one channel for one motor. Timer1 is set on Phase Correct PWM mode.

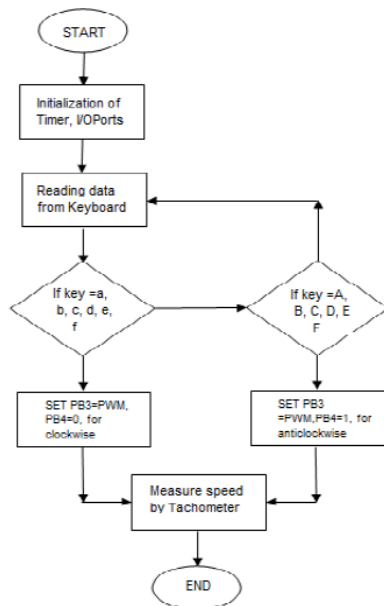


Figure. 2. Algorithm for Speed control

Timer 2 in CTC Mode was used to generate constant time periods. Programming tools used for the proposed work are embedded C – to generate PWM for the H-bridge switches using microcontroller and MATLAB software package where graphical programming is done to provide Graphical User Interface.

Fast PWM Mode Duty Cycle

Duty cycle can be calculated by using equation 1. The duty cycle can be determined using the OCR0 register, bigger OCR0 value results in a bigger duty cycle. When OCR0 is 255, the OC0 is 256 clocks out of 256 clocks, which means duty cycle is 100percent.

$$DUTY\ CYCLE = \frac{OCR0+1}{256} \times 100 \quad (1)$$

SPEED CONTROL:

DC motor converts electrical energy into mechanical energy. DC motor is used in applications where wide speed

ranges are required. DC shunt motor exhibits a drooping speed-torque characteristic. The speed of the DC motor is given by

$$N = \frac{V - I_a R_a}{Z\phi} \cdot \left(\frac{A}{P}\right) = k \cdot \left(\frac{V - I_a R_a}{\phi}\right)$$

Hence, the speed can be controlled by varying,

- i) Flux/pole (ϕ). Field Control Method
- ii) Voltage of armature circuit, by varying R_a , Armature Control Method.

PROPOSED FUZZY LOGIC CONTROLLER:

In most of the adaptive fuzzy controllers, attempt is made to change the Rule Base to make the system adaptive. In the scheme proposed, such a result is achieved by adjusting the defuzzifier as a function of the system response. Also it is possible to regulate the parameters of the time – domain response. The block diagram of the system with the proposed FLC is shown in Fig 5. A new functional block called the Error Interpreter is added to the basic system of Fig 2. The function of the block is to sense the error, identify its ranges. And determine the location of the singletons.

In this method, the error and error rate are used to change the supporters in the motor voltage singletons. The defuzzified output of controller is given by

$$V_c = \frac{PLVC.PL + PSVC.PS + NSVC.NS + NLVC.NL}{PL + PS + NS + NL}$$

Where $PL, PS, NS,$ And NL is the inference membership values and $PLVC, PSVC, NSVC,$ and $NLVC$ are the corresponding supports of change in the motor voltage singletons. In the new method, the error signal is fed to an interface that changes the value of the supports.

The magnitude of output error is divided into ranges covering (100 – 40), (40 - 20), (20 - 5), and (5 - 0) percent of the maximum output. For a lower range of error, the supports are multiplied by a coefficient less than unity (around set point) and for higher ranges, the supports are multiplied by a coefficient greater than unity. The following method is suggested for controlling the time – response parameters:

Range 1: (100 – 40). This range is used to effectively control the rise time and to obtain maximum overshoot. If the coefficient of supports is increased, the rise time is decreased and vice versa.

Range 2: (40 - 20). The variation of the coefficient during this range will affect the maximum overshoot by about 80% and the variation in each of ranges 1 and 3 will affect by about 10%.

Range 3 and 4: (20 – 5) (5 – 0). The coefficient of this range has the maximum effect on the steady state oscillations. If the coefficients are larger, the oscillations will persist for a longer time and thus the setting time will be more.

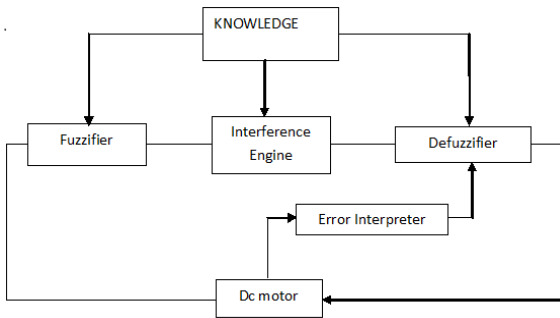


Fig.3. Block diagram of system with proposed FLC

The details of the above symbols are as follows:

V_a – the applied armature voltage

L_a – the armature inductance

R_a – the armature resistance

E_b – the back e.m.f

K – the back e.m.f. or torque constant

ω – the angular speed of the motor

T_m – the developed motor torque

J – the moment of inertia of the motor including load

B – the viscous co-efficient

T_l – the applied load torque.

The steps in designing the controller are:

- 1) Identify the variables (inputs, states and outputs of the) of the plant.
- 2) Partition the universe of discourse or the internal spanned by each variable into a number of fuzzy subsets, assigning each a linguistic label.
- 3) Assign or determine a membership function for each fuzzy subset.
- 4) Assign the fuzzy relationship between the inputs or states, fuzzy subsets on the one hand and the outputs fuzzy subsets on the other hand, thus forming the rule base.
- 5) Choose appropriate scaling factors for the input and output variables in order to normalize the variables to the [0,1] or [-1,1] interval.
- 6) Fuzzify the inputs to the controller.
- 7) Use fuzzy approximate reasoning to infer the output contributed from each rule.
- 8) Aggregate the fuzzy outputs recommended by each rule.
- 9) Apply defuzzification to form a crisp output.

Thus based upon these rules fuzzy logic controller is designed and can be suitable for any kind of control applications.

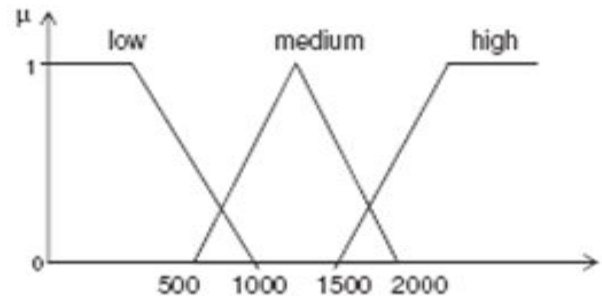


Fig 4. Illustrates Membership for Fuzzy Variable “speed” in R.P.M

	NL	NS	ZERO	PS	PL
PL		PL		NL	NL
PS	PS	ZERO	NS	NL	NL
ZERO	PL	PS	ZERO	NS	NL
NS	PL	PL	PS	ZERO	NS
NL	PL	PL		PS	

Fuzzy control rules

IV. MODELING OF PERMANENT MAGNET DC MOTOR

PROPOSED MODEL

To develop a model of DC motor speed control using Matlab/Simulink the following equations are used for the modeling. The equations of armature circuit

$$V_a = L_a \frac{di_a}{dt} + R_a i_a + E_b \quad (1); \quad E_b = K \omega \quad (2)$$

The equations of the mechanical side

$$T_m = K i_a \quad (3); \quad T_m = J \frac{d\omega}{dt} + B\omega + T_l \quad (4)$$

V. SIMULATION RESULTS

(i) OPEN LOOP CONTROL

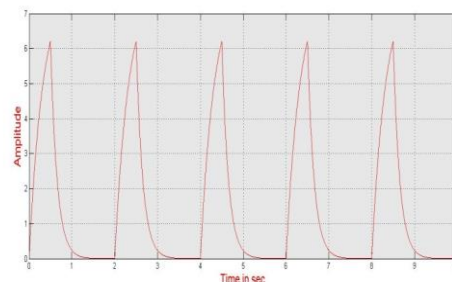


Fig.5. Transient speed response of the microcontroller generated PWM driven motor under open loop condition

(ii) CLOSED LOOP CONTROL

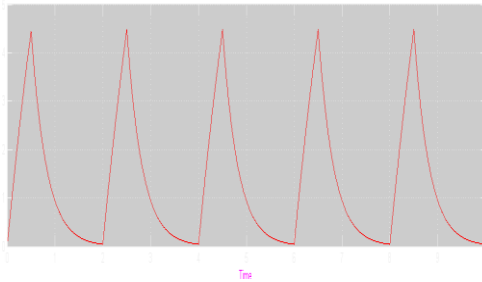


Fig.6. Pulse Width Modulation based DC motor controller

(iii) FUZZY LOGIC CONTROL

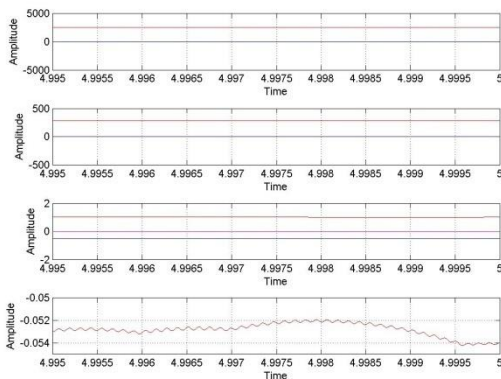


Fig7 . Fuzzy logic based DC motor controller

VI. CONCLUSION

In this paper Fuzzy controller has more design parameters, based on empirical software rules, more suitable to satisfy non-linear criterion in all operation range compared with PID controller (satisfy linear criterion), which is based on a hardware components. In conventional PID controller design, mathematical model of the system must be derived then a mathematical model of a controller could be developed. But, in fuzzy logic controller design, no need for mathematical representation of the system because it depends basically on human experience, hence, its easier, to design the controller to such system. In fuzzy logic controller design for speed control of D.C motors by controlling on firing angle value of the bridge converter, also the fuzzy logic controller takes the difference between the reference rotated speed and the actual motor speed and then gives appropriate firing angle to reduce the first error between the reference and the actual rotated speed. In PID controller the armature voltage of a separately excited D.C motor is varying between (230 - 265) V, while in fuzzy logic controller is varying between (145 - 265) V.

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