# Design and Simulation of Double Precision Floating-Point Adder 

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#### Abstract

Floating point numbers are very important part of computer processing. Addition imposes a great challenge due to its processing time. This paper presents a technique for addition of IEEE 754 double precision floating-point numbers within two clock cycles. This paper results also show improvements in power utilization, operational chip area management and optimization of hardware. This proposed adder is implemented with the help of 6slx45tfgg484-3 Spartan family as well as 5vfx70tff1136-1 of Virtex Xilinx FPGA devices.


Keywords-IEEE 754,Floating Point Addition,Adder,FPGA

## I. Introduction

By using floating point numbers in computation we can represent number in a way that it can tradeoff between range as well as precision. As we know computer memory is limited so we cannot store any number with infinite precision so we use floating point numbers. From early times so many formats have been used but since 1990 IEEE make a standard for floating point numbers (IEEE 754).In this there are two kinds of format IEEE 754 single precision and IEEE 754 double precision format. Floating point numbers are represented in the form.

| Sign Bit | Biased exponent | Significand/mantissa/fraction |
| :--- | :--- | :--- |

Sign bit represents the sign of the number if it is zero then it represents a positive number and if it is zero then represents negative number, exponent field determine range of numbers which can be represented, significand represents precision of number.

| PARAMETER | IEEE-754 SINGLE <br> PRECISION | IEEE-754 DOUBLE <br> PRECISION |
| :--- | :---: | :---: |
| Total no. of bits | 32 | 64 |
| Significand bits | $23+1$ | $52+1$ |
| Exponent bits | 8 | 11 |
| Sign bit | 1 | 1 |
| Special e value | 255 | 2047 |
| Bias | 127 | 1023 |

Here we concentrate on double precision floating point numbers. Simply if we want to convert a number into IEEE 754 double precision format. First we need to convert that number into binary then it should be normalized and exponent is biased and then written in standard format. Suppose we have number (-3.75). Then binary conversion of 3 is (11) and for 0.75 it is (1011).So (3.75) is equal to ( $1.11011 * 2^{\wedge} 1$ ), as exponent is one here it must be biased means $(1023+1)=1024$ it is again converted into binary. The final conversion of (3.75) into IEEE 754 double precision format is 1 10000000000111000000000000000000000000000000000000 000000000000000.

Lots of work have been done in this field. First implementation is done by l.louca and T.Cook and W.Johnson in 1996.The algorithm in[2] implement adder, subtractor and multiplier designs and achieved the operating frequencies of 363.76 Mhz and 414.714 hz with an area of 660 and 648 slices. Error detection technique is presented in [3].Optimization of each individual component of adder has been done in [4].Overall speed is the main concern of the operation due to large number of bits.

The proposed algorithm is implemented with the help of Xilinx FPGA device 6slx45tfgg484-3 and 5vfx70tff11361.Various parameters like number of slice flip-flops, Number of 4 input LUTs, number of global clks have been observed. Proposed technique show improvements in hardware optimization, latency and chip area compare to algorithm used in [1].

Rest of the paper is organized as section II presents addition of two floating point numbers. Section III describes the proposed algorithm. Section IV shows the device usage section V shows the simulation results. Section VI contain conclusion and future scope is provided in section VII.

## II. ADDITION OF TWO IEEE 754 DOUBLE PRECISION FLOATING POINT NUMBERS

In order to add to given numbers into double precision the two numbers primarily converted into double precision format and the exponents of two numbers are compared if it is equal then we can directly add the numbers and normalize the answer but if the exponent is not equal then we need to rewrite smaller number such that its exponent matches with the larger number and then addition has been performed. After the result of addition it must be normalized.


## III. Proposed algorithm

Proposed algorithm is similar to as used in [1].The floating point arithmetic operation uses two clock cycles here. In [1] it is also a two staged pipeline which is divided into two paths. The two paths ("R path" and "N path") are selected on the basis of the exponent difference. The two pipeline stages execute in two different clock cycles. In very first step for this paper, algorithm is written for determining the exponent

difference, then big number is determined out of two exponents by implementing conditions when both the exponents have same sign and both the exponents have different sign. Then greater mantissa is selected for shifting operation and shifting is done for equalization of exponents. All of this operation is done in $1^{\text {st }}$ clock cycle. In next clock cycle two mantissa are added and then normalization has been done.

## IV. IMPLEMENTATION DETAILS

Somsubhra Ghosh implemented the proposed algorithm with XC2V6000 and XC3S1500 xilinx FPGA devices[1].Similar algorithm when implemented with 6s1x45tfgg484-3 improvement in results has been noticed and tabulated as follows.

Table1: Estimation of usage of resources in device 6 slx 45 tfgg 484-3

| Device Utilization Summary |  |  |  |
| :--- | :---: | :---: | :---: |
| Logic Utilization |  |  |  |
| Slice Logic Utilization | Available | Utilization |  |
| Number of Slice Registers | 63 | 54576 | $1 \%$ |
| Number of Slice LUTs | 645 | 27288 | $2 \%$ |
| Number used as Logic | 645 | 27288 | $2 \%$ |
| Slice Logic Distribution: |  |  |  |
| Number of LUT Flip Flop pairs used | 645 |  |  |
| Number with an unused Flip Flop | 582 | 645 | $90 \%$ |
| Number with an unused LUT | 0 | 645 | $0 \%$ |
| Number of fully used LUT-FF pairs | 63 | 645 | $7 \%$ |
| IO Utilization |  |  |  |
| Number of IOs | 193 |  |  |
| Number of bonded IOBs | 192 | 296 | $64 \%$ |

Minimum input arrival time before clock is: 27.424 ns Maximum output required time after clock: 5.463ns Maximum combinational path delay: 39.551 ns
Latency: 2 clock cycles
Table 2: Estimation of usage of resources in device 5vfx70tff136-1

| Device Utilization Summary |  |  |  |
| :---: | :---: | :---: | :---: |
| Logic Utilization | Used | Available | Utilization |
| Slice Logic Utilization |  |  |  |
| Number of Slice Registers | 63 | 44800 | 0\% |
| Number of Slice LUTs | 602 | 44800 | 1\% |
| Number used as Logic | 602 | 44800 | 1\% |
| Slice Logic Distribution: |  |  |  |
| Number of LUT Flip Flop pairs used | 602 |  |  |
| Number with an unused Flip Flop | 539 | 602 | 89\% |
| Number with an unused LUT | 0 | 602 | 0\% |
| Number of fully used LUT-FF pairs | 63 | 602 | 10\% |
| IO Utilization |  |  |  |
| Number of IOs | 193 |  |  |
| Number of bonded IOBs | 192 | 640 | 30\% |

Table 2 give results of same algorithm on virtex 5 .Number of resources are increased from Spartan 6 to virtex 5 and same results are arrived this simply proves that when same algorithm implemented on virtex 2 we get improved results.

## V. SIMULATION RESULTS



## VI. CONCLUSION

The double precision floating point adder is successfully implemented .If the algorithm is implemented on the advanced versions then better results will be derived in future.

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